

Power Analysis :

Instantaneous power = $p(t) = i(t) v(t) = I_m \cos(\omega t + \theta_i) V_m \cos(\omega t + \theta_v)$
(lower case) AC currents

time-independent parameters: $\left\{ \begin{array}{l} I_m \text{ \& } V_m \text{ are magnitudes of alternating current and voltage, respectively} \\ \theta_i \text{ \& } \theta_v \text{ are the respective phase angles.} \end{array} \right.$

Average power = P (upper case) $= \frac{1}{T} \int_{t_0}^{t_0+T} p(t) dt$
(our current & voltage are periodic w/period $T = \frac{2\pi}{\omega}$)

$$P = \frac{1}{T} \int_T \frac{1}{2} I_m V_m \{ \cos(\cancel{\omega t} + \theta_i - \cancel{\omega t} - \theta_v) + \cos(\omega t + \theta_i + \omega t + \theta_v) \} dt$$

Trig. identity: $\cos \alpha \cos \beta = \frac{1}{2} \{ \cos(\alpha - \beta) + \cos(\alpha + \beta) \}$

$$P = \frac{1}{T} \int_T dt \underbrace{\frac{1}{2} I_m V_m \cos(\theta_i - \theta_v)}_{\text{time-independent}} + \frac{1}{T} \int_T dt \frac{1}{2} I_m V_m \cos(2\omega t + \theta_i + \theta_v)$$
$$= \underbrace{\frac{1}{2} I_m V_m \cos(\theta_i - \theta_v)} + \frac{1}{2} I_m V_m \underbrace{\frac{1}{T} \int_T dt \cos(2\omega t + \theta_i + \theta_v)}_{\substack{\downarrow \text{period: is } \frac{T}{2} \\ \text{average over 2 periods of a sinusoid} \rightarrow 0}}$$

$P = \frac{1}{2} I_m V_m \cos(\theta_i - \theta_v)$

Makes sense!

At a pure-reactive element = $P = 0$
(inductor or capacitor)
At a pure-resistive element: $P = \frac{1}{2} I_m V_m$

at an inductor:
(~~charge~~ magnetic energy storage)

$$\hat{Z}_L = j\omega L = \frac{\hat{V}}{\hat{I}} = \frac{V_m}{I_m} \angle \theta_v - \theta_i$$

$$90^\circ = \theta_v - \theta_i \rightarrow \cos(\theta_v - \theta_i) = 0 \quad P = 0$$

at a capacitor:
(electric energy storage)

$$\hat{Z}_C = \frac{1}{j\omega C} = \frac{\hat{V}}{\hat{I}} = \frac{V_m}{I_m} \angle \theta_v - \theta_i$$

$$-90^\circ = \theta_v - \theta_i \rightarrow \cos(\theta_v - \theta_i) = 0 \quad P = 0$$

RMS (root-mean-square) or effective values:

$$\text{Current: } i(t) = I_m \cos(\omega t + \theta_i)$$

$$\text{RMS value or effective value is } I_{RMS} = \sqrt{\frac{1}{T} \int_T i^2(t) dt}$$

$$= I_m \sqrt{\frac{1}{T} \int_T dt \cos^2(\omega t + \theta_i)} = \frac{I_m}{\sqrt{2}} \sqrt{\frac{1}{T} \int_T dt (1 + \cos(2\omega t + 2\theta_i))}$$

Trig. id: $\cos^2 \alpha = \frac{1}{2} (1 + \cos 2\alpha)$

$$= \frac{I_m}{\sqrt{2}} \left(\sqrt{1 + \underbrace{\frac{1}{T} \int_T \cos(2\omega t + 2\theta_i) dt}_{\text{average over 2 periods of a sinusoid} \Rightarrow 0}} \right) \Rightarrow \boxed{I_{RMS} = \frac{I_m}{\sqrt{2}}}$$

$$\text{Voltage: } v(t) = V_m \cos(\omega t + \theta_v) \rightarrow \boxed{V_{RMS} = \frac{V_m}{\sqrt{2}}}$$

$$\rightarrow P = \frac{1}{2} I_m V_m \cos(\theta_i - \theta_v) = I_{RMS} V_{RMS} \cos(\theta_i - \theta_v)$$

$$\boxed{P = I_{RMS} V_{RMS} pf} \left\{ \begin{array}{l} \text{at pure-reactive: } 0 \\ \text{at pure-resistive: } I_{RMS} V_{RMS} \end{array} \right.$$

≡ power factor
or p.f.

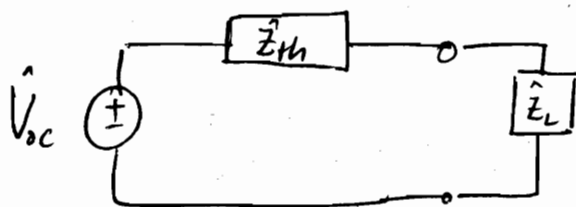
Complex Power = (not measurable, just a definition)

$$\begin{aligned}\hat{S} &= \hat{V}_{RMS} \hat{I}_{RMS}^* = V_{RMS} \angle \theta_v \cdot I_{RMS} \angle -\theta_i \\ (\text{upper case}) & \\ &= V_{RMS} I_{RMS} \angle (\theta_v - \theta_i) \quad \text{polar form} \\ &= V_{RMS} I_{RMS} \cos(\theta_v - \theta_i) + j V_{RMS} I_{RMS} \sin(\theta_v - \theta_i) \quad \text{rectangular form}\end{aligned}$$

$$= \underbrace{(\underline{P})}_{\uparrow} + j V_{RMS} I_{RMS} \sin(\theta_v - \theta_i)$$

The real part of the complex power $\hat{S}$ is the average power $\underline{P}$.

→ Given a circuit, what load would draw the maximum power?
To find out, let's replace our circuit by a Thevenin equivalent.



What $\hat{Z}_L$ would extract the max. power? It's $\hat{Z}_L = \hat{Z}_{th}^*$

Proof: Ave. power at load is $\underline{P}_L = \frac{1}{2} I_L V_L \cos(\theta_{V_L} - \theta_{I_L})$

$$= \frac{1}{2} \frac{V_{oc}^2 (R_L^2 + X_L^2)^{1/2}}{(R_{th} + R_L)^2 + (X_{th} + X_L)^2} \frac{R_L}{(R_L^2 + X_L^2)^{1/2}}$$

To get max. P_L : 1) Get smaller denominator: $X_L = -X_{th}$

$$\Rightarrow P_L = \frac{1}{2} \frac{V_{oc}^2 R_L}{(R_L + R_{th})^2}$$

2) Max P_L w.r.t. R_L :

$$\frac{\partial P_L}{\partial R_L} = 0$$

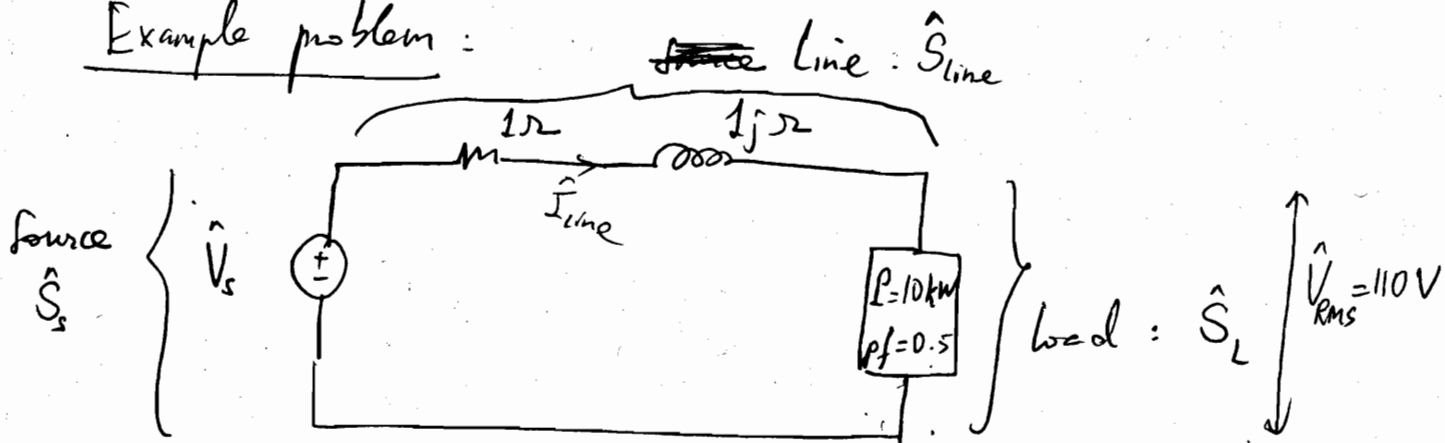
$$\frac{1}{2} \frac{V_{oc}^2}{(R_{th} + R_L)^2} - \frac{V_{oc}^2 R_L}{(R_{th} + R_L)^3} = 0$$

$$\frac{1}{2} - \frac{R_L}{R_{th} + R_L} = 0 \rightarrow 2R_L = R_{th} + R_L$$

$$\Rightarrow \boxed{R_L = R_{th}}$$

$$\Rightarrow \boxed{\hat{Z}_L = \hat{Z}_{th}^*}$$

Example problem :

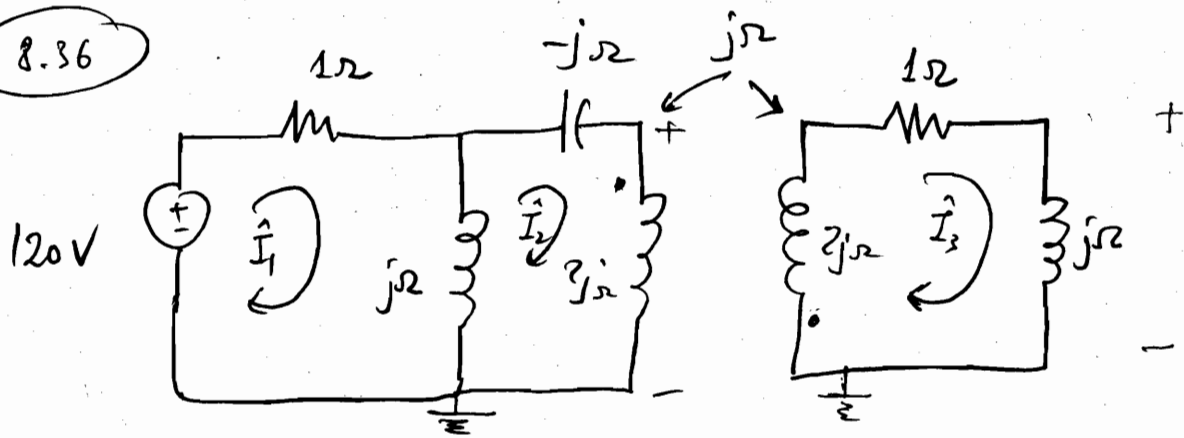


Find $\hat{S}_s$ and p.f at the source. (To design a power supply)

$$\hat{S}_s = \hat{S}_{line} + \hat{S}_L$$

4) $\hat{S}_L$: $\rightarrow$ cont. on page 48

8.36

Loop analysis:

$$1) \quad 120 = I_1 + j(I_1 - I_2) \quad \checkmark = I_1(1+j) - jI_2$$

$$2) \quad 0 = (I_2 - I_1)j - jI_2 + jI_2 \pm jI_3$$

$$= -jI_1 + jI_2 \pm jI_3$$

$$3) \quad 0 = jI_3 + jI_2 + (1+j)I_3$$

$$= (1+3j)I_3 + jI_2$$

$$\hookrightarrow I_3 = \frac{-j}{1+3j} I_2$$

$$0 = -jI_1 + I_2 \left(j - \frac{j^2}{1+3j} \right)$$

$$I_2 = \frac{jI_1}{j - \frac{j^2}{1+3j}}$$

$$\frac{1}{Z_1} = \frac{1}{1+j}$$

$$120 = I_1(1+j) + I_1 \frac{1}{j + \frac{1}{1+3j}}$$

$$= I_1 \left[1+j + \frac{1}{j + \frac{1}{1+3j}} \right]$$

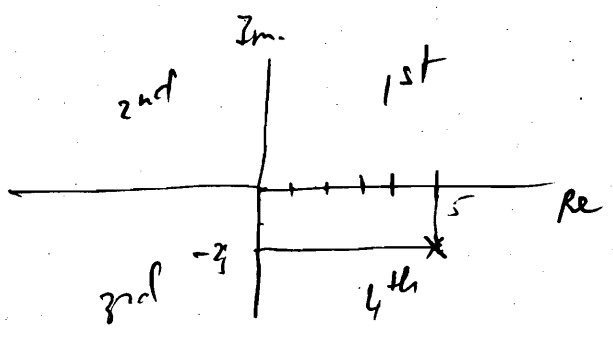
$$120 = \hat{I}_1 \left[1+j + \frac{1+3j}{2j-6+1} \right] = \hat{I}_1 \frac{(1+j)(-5+4j) + 1+3j}{-5+2j}$$

$$= \hat{I}_1 \frac{-7 - 3j + 1 + 3j}{-5+2j} = \hat{I}_1 \frac{6}{5-2j}$$

$$\hat{I}_1 = 20(5-2j)$$

$$\rightarrow \hat{Z}_0 = \frac{120}{\hat{I}_1} = \frac{120}{20(5-2j)} = \frac{6}{5-2j} = \frac{6}{\sqrt{29} \angle -\tan^{-1} \frac{2}{5}}$$

$$= 1.11 \angle 21.8^\circ$$



Solution answer is correct but sign for mutual inductance term in loop #2 should be +.

$$\angle \theta_v - \theta_i = \cos^{-1}(p.f.) = \cos^{-1}(0.5) = 60^\circ$$

(42)

a) $\hat{S}_L = \underbrace{V_{RMS} I_{RMS}}$

We know $P = 10 \text{ kW}$; $p.f. = 0.5 \rightarrow V_{RMS} I_{RMS} ?$

Since $\text{Re}[\hat{S}_L] = P = V_{RMS} V_{RMS} p.f. \rightarrow V_{RMS} I_{RMS} = \frac{P}{p.f.} = \frac{10 \text{ kW}}{0.5} = 20 \text{ kW}$

$\hat{S}_L = 20000 \angle 60^\circ \text{ V.A}$
same as W

b) $\hat{S}_{\text{line}} = \hat{V}_{\text{line}} \hat{I}_{\text{line}}^* = \underbrace{\left(\hat{I}_{\text{line}} \hat{Z}_{\text{line}} \right)}_{\text{Ohm's law}} \hat{I}_{\text{line}}^* = I_{\text{line}}^2 \hat{Z}_{\text{line}}$
def. of complex power

$= I_{\text{line}}^2 (1+j)$

$\hat{I}_{\text{line}} = \hat{I}_L \text{ (series)} \rightarrow \hat{I}_L = \left(\frac{\hat{S}_L}{\hat{V}_L} \right)^* = \left(\frac{20000 \angle 60^\circ}{110 \angle 0^\circ} \right)^* = 181.82 \angle -60^\circ \text{ A}$

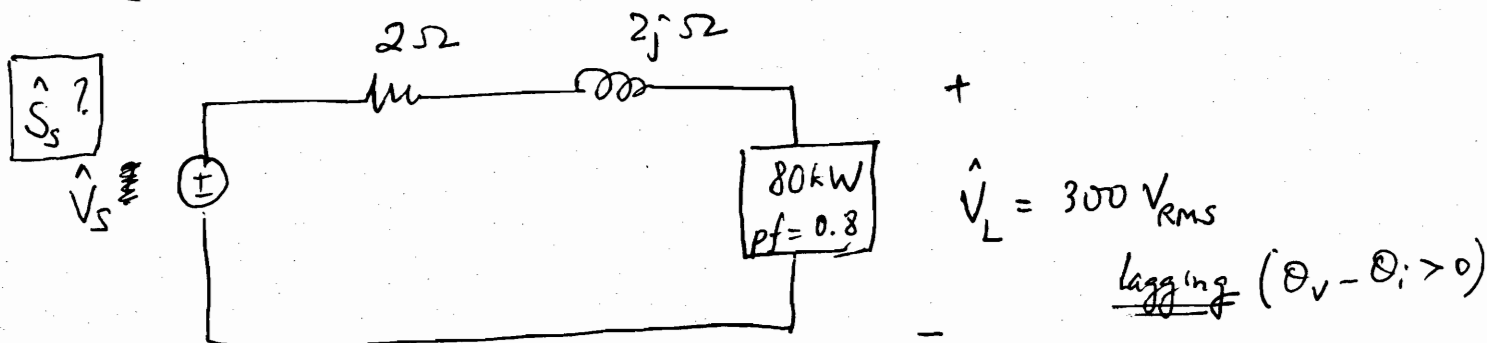
$\hat{S}_{\text{line}} = 181.82^2 (1+j) \text{ V.A}$

c) $\hat{S}_S = \hat{S}_{\text{line}} + \hat{S}_L = 181.82^2 (1+j) + 20000 (\cos 60 + j \sin 60)$
 $= 43058.5 + j 50379 \text{ VA} = 66272.8 \angle 49.5^\circ \text{ VA}$

p.f. at source = $\cos$ of the angle of $\hat{S}_S = \cos(49.5^\circ) = 0.65$

d) $\hat{V}_S = \frac{\hat{S}_S}{\hat{I}_L^*} = \frac{66272.8 \angle 49.5^\circ}{181.82 \angle 60^\circ} = 364.49 \angle -10.5^\circ \text{ V}$
 $\hat{S}_S = \hat{V}_S \hat{I}_S^* = \hat{V}_S \hat{I}_L^*$
series

Quiz #2 March 27, 2007



Find the complex power at the source $\hat{S}_s$.

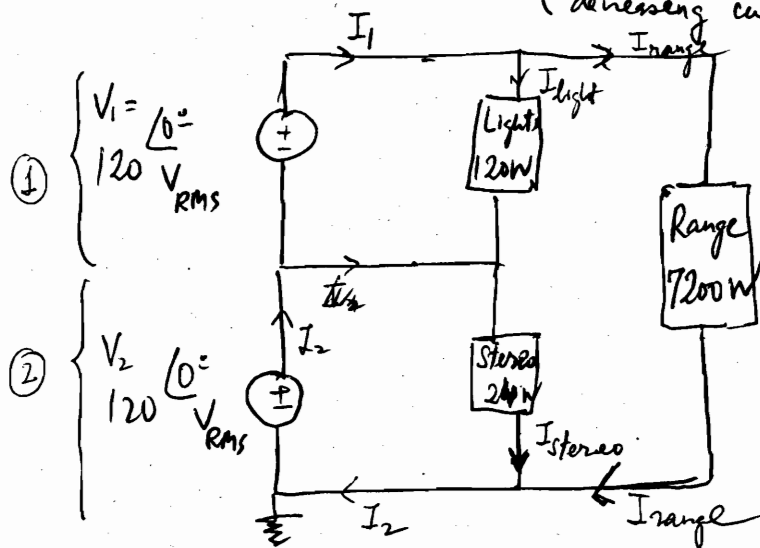
HW6 due 4/3/07

HW7 due 4/10/07

Exam 2 4/12/07 (HW's 4-7)

Single-phase three-wire circuit:

increase voltage difference for stove and dryer
(decreasing current required)



- Single-phase of 0°
- Three-wire

Utility bill per month?
\$0.08/kWh

Total energy consumed a day (lights on 15 hours, Range on 4 hours, stereo on 8 hours)

Given Power and time $\rightarrow$ energy consumed is $P \times \text{time}$.

Energy supplied by source #1 is

$$\begin{aligned}
 E_1 &= \int P_1 dt = \int V_1 I_1 dt = V_1 (I_1 \times \text{time}) \\
 &= 120V \left[\underbrace{I_{\text{lights}} \times 15 \text{ hrs}}_{\frac{120W}{120V}} + \underbrace{I_{\text{Range}} \times 4 \text{ hrs}}_{\frac{7200W}{240V}} \right] = \overset{120 \times}{15 \text{ Whrs}} + \overset{120 \times}{14400 \text{ Wh}} \\
 &= 14.4 \text{ kWh} + 1.8 \text{ kWh} = 16.2 \text{ kWh}.
 \end{aligned}$$

Energy supplied by source #2 is

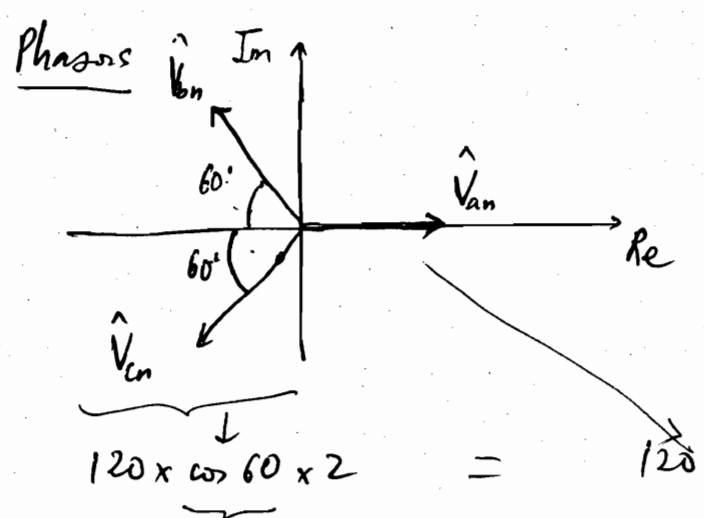
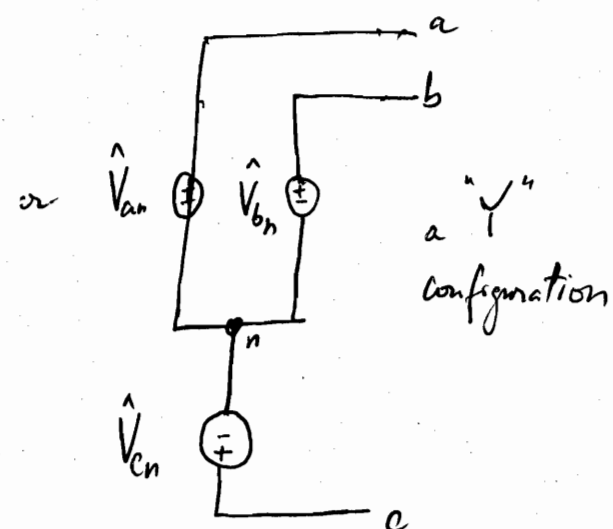
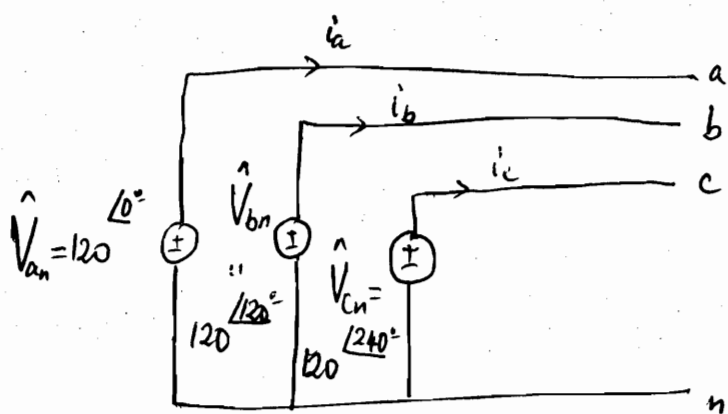
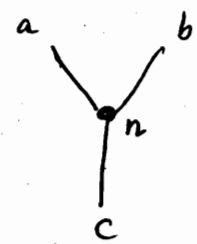
$$\begin{aligned}
 E_2 &= \int P_2 dt = \int V_2 I_2 dt = V_2 (I_2 \times \text{time}) \\
 &= 120V \times \left[\underbrace{I_{\text{stereo}} \times 8 \text{ h}}_{\frac{24W}{120V}} + \underbrace{I_{\text{Range}} \times 4 \text{ h}}_{\frac{7200W}{240V}} \right] = 192 \text{ Wh} + 14,400 \text{ Wh} \\
 &= 14.592 \text{ kWh}
 \end{aligned}$$

$$\begin{aligned}
 \text{Total energy consumed} &= \text{total energy delivered} = E_1 + E_2 \\
 &= (16.2 + 14.592) \text{ kWh} = 30.8 \text{ kWh} \\
 &\quad \text{(a day)}.
 \end{aligned}$$

$$\rightarrow \text{Electric cost a month} = 30.8 \text{ kWh} \times 30 \times \frac{\$0.08}{\text{kWh}} = \$73.9 \text{ a month}$$

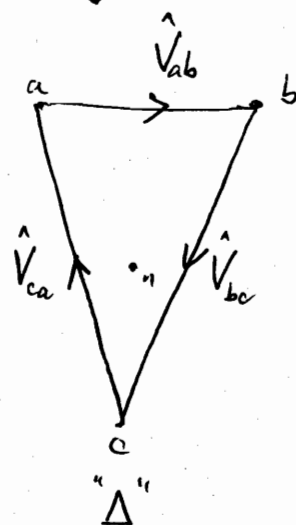
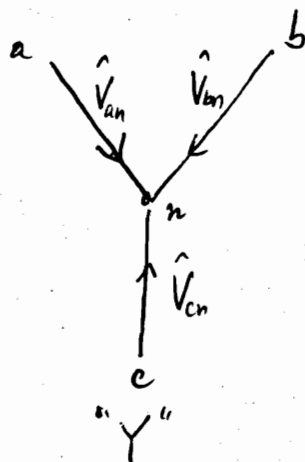
Three - Phase Circuits =

"Y" configuration



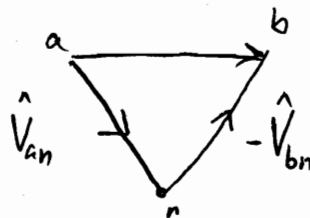
$\hat{V}_{an} + \hat{V}_{bn} + \hat{V}_{cn} = 0 \rightarrow$ "Balanced three phase circuit"
 Ordered in CCW direction $\rightarrow$ "~~abc~~ balanced three-phase circuit"
acb

" Δ " or delta configuration :



Through geometrical consideration, we can go from "Y" voltages to " Δ " voltages and viceversa

$$\hat{V}_{ab} = \hat{V}_{an} - \hat{V}_{bn}$$



rectangular form

$$= V_p - (V_p \cos 120 + j V_p \sin 120)$$

$$(V_p = 120) = V_p (1 - \cos 120) - j V_p \sin 120 = V_p \left(\frac{3}{2} - j \frac{\sqrt{3}}{2} \right)$$

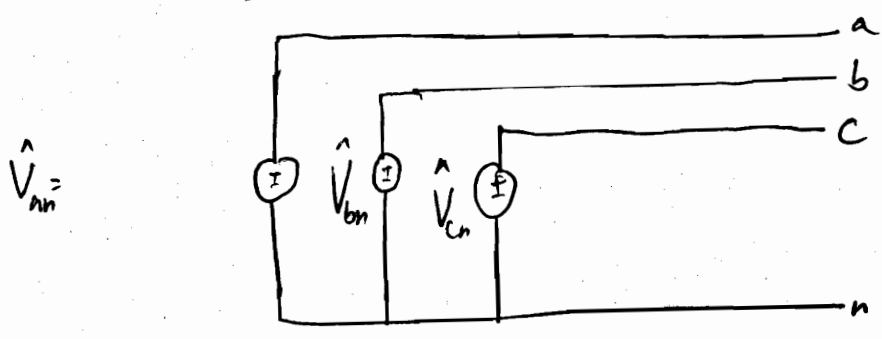
polar form

$$= V_p \sqrt{\frac{9}{4} + \frac{3}{4}} \angle \tan^{-1} \left(\frac{-1}{\sqrt{3}} \right) = V_p \sqrt{3} \angle -30^\circ$$

$$\begin{aligned}\hat{V}_{bc} &= \hat{V}_{bn} - \hat{V}_{cn} = V_p (\cos 120^\circ + j \sin 120^\circ) - V_p (\cos (-120^\circ) + j \sin (-120^\circ)) \\ &= V_p j \underbrace{\frac{\sin(120)}{2} \times 2}_{\frac{\sqrt{3}}{2}} = V_p \sqrt{3} j \\ &= V_p \sqrt{3} \angle 90^\circ\end{aligned}$$

$$\begin{aligned}\hat{V}_{ca} &= \hat{V}_{cn} - \hat{V}_{an} = V_p (\cos (-120^\circ) + j \sin (-120^\circ)) - V_p \\ &= V_p (\cos 120 - 1) - j V_p \sin 120 = V_p \underbrace{\left(-\frac{3}{2} - j \frac{\sqrt{3}}{2}\right)}_{\text{3rd quad.}} \\ &= V_p \sqrt{3} \angle \tan^{-1} \frac{1}{\sqrt{3}} + 180^\circ \\ &= V_p \sqrt{3} \angle 210^\circ\end{aligned}$$

"ABC balanced three-phase circuit":



$$\begin{aligned}\hat{V}_{an} &= 120 \angle 0^\circ = V_p \angle 0^\circ \\ \hat{V}_{bn} &= 120 \angle 240^\circ = V_p \angle -120^\circ \\ \hat{V}_{cn} &= 120 \angle 120^\circ\end{aligned}$$

Phasors:

